

Mathematical Foundation I: Fourier Transform, Bandwidth, and Band-pass Signal Representation

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Fourier Transform

- $g(t)$: a non-periodic deterministic signal.
- Definition: the Fourier transform of the signal $g(t)$ is

$$G(f) = \int_{-\infty}^{\infty} g(t) \exp(-j2\pi ft) dt$$

- $j = \sqrt{-1}$
- t : time (in second)
- f : frequency (in Hz)
- Definition: $\exp(i\theta) = \cos(\theta) + i \sin(\theta)$

Fourier Transform

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- Definition: the Fourier transform of the signal $g(t)$ is

$$G(f) = \int_{-\infty}^{\infty} g(t) \exp(-j2\pi ft) dt$$

- or we can have $\omega = 2\pi f$
- and it becomes

$$G(f) = \int_{-\infty}^{\infty} g(t) \exp(-j\omega t) dt$$

Inverse Fourier Transform

- The original signal can be recovered exactly using inverse Fourier Transform:

$$g(t) = \int_{-\infty}^{\infty} G(f) \exp(j2\pi ft) df$$

- We call $g(t) \Leftrightarrow G(f)$ a Fourier transform pair.
- We can write $G(f) = F[g(t)]$ or $g(t) = F^{-1}[G(f)]$

The meaning of Fourier transform

$$g(t) = \int_{-\infty}^{\infty} \underbrace{G(f) \exp(j2\pi ft)}_{\text{Contribution of one frequency to } g(t)} df$$

Summing over all frequencies

Weighted by $G(f)$

Sinusoids at different frequency f

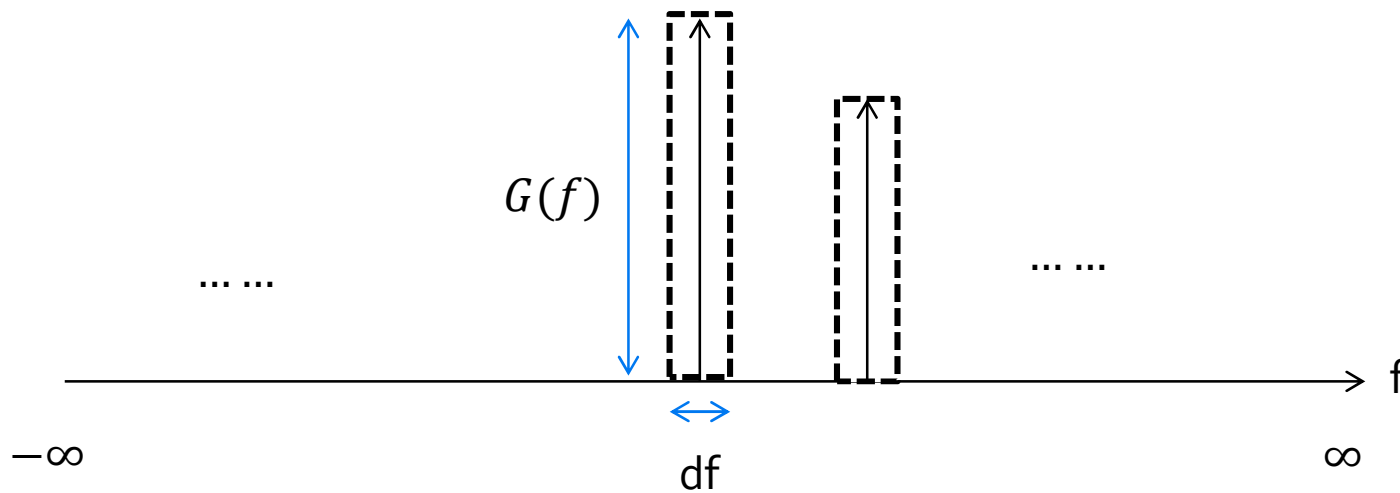
infinitesimally small width

Decomposition of Rectangular Waves using sinusoids:
http://www.youtube.com/watch?v=y6crWlxKB_E

The meaning of Fourier transform

Fourier transform:

A tool to resolve a given signal $g(t)$ into its complex exponential components occupying the entire frequency interval from $-\infty$ to ∞ .



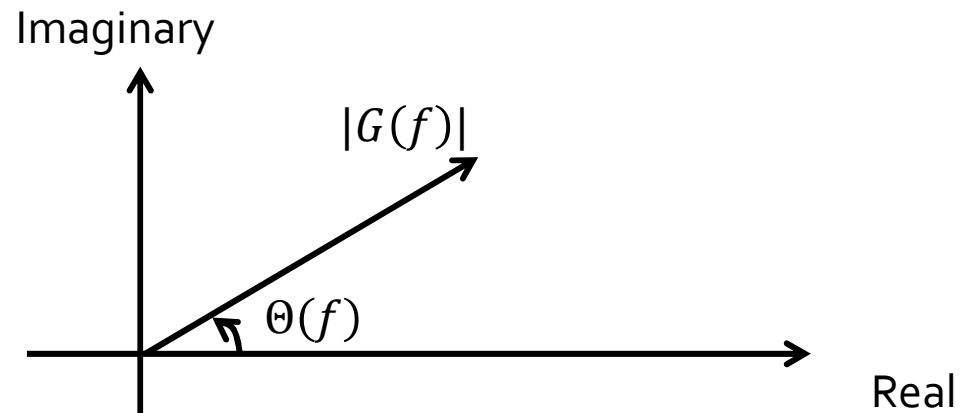
Continuous Spectrum

- $G(f)$ is a complex function of frequency f , so it can be expressed as:

$$G(f) = |G(f)| \exp(j\theta(f))$$

Continuous amplitude spectrum

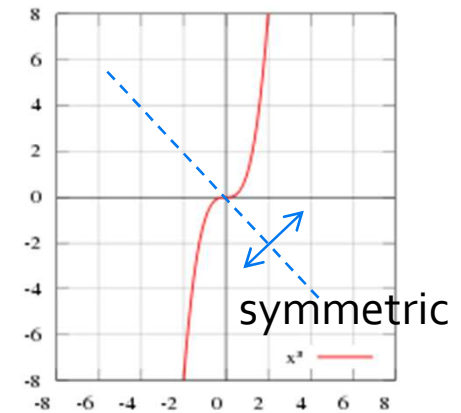
Continuous phase spectrum



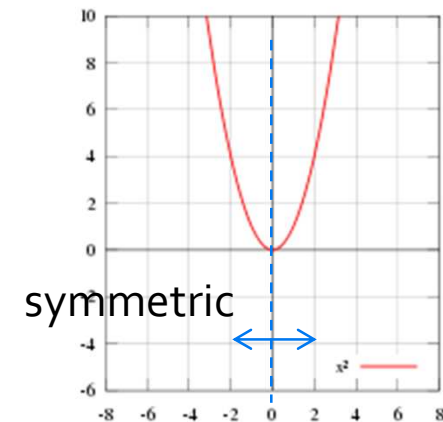
Real-value signal

- **Conjugation:**
if $G(f) = a + bj$, then $G^*(f) = a - bj$
- **Odd function:** $f(-x) = -f(x)$
- **Even function:** $f(x) = f(-x)$
- **For a real $g(t)$:** $G(-f) = G^*(f)$
- **And**
 - $|G(-f)| = |G(f)|$ ← even function
 - $\theta(-f) = -\theta(f)$ ← odd function

odd function



even function

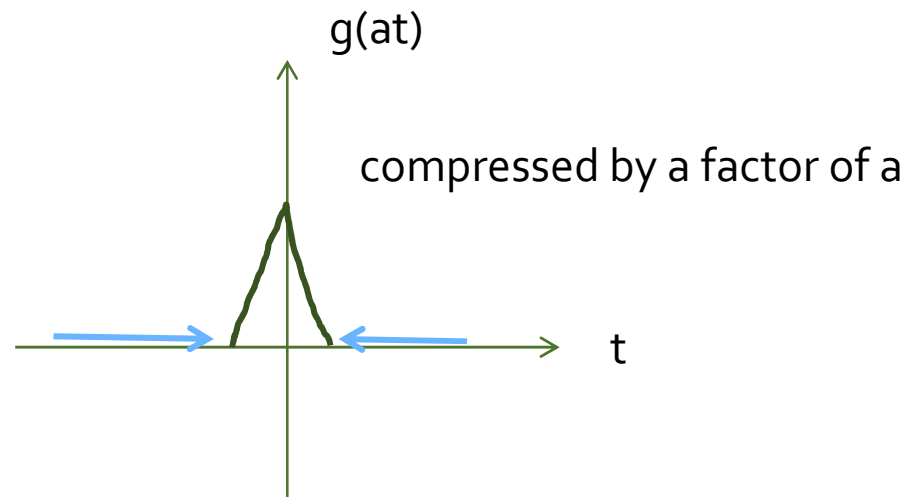
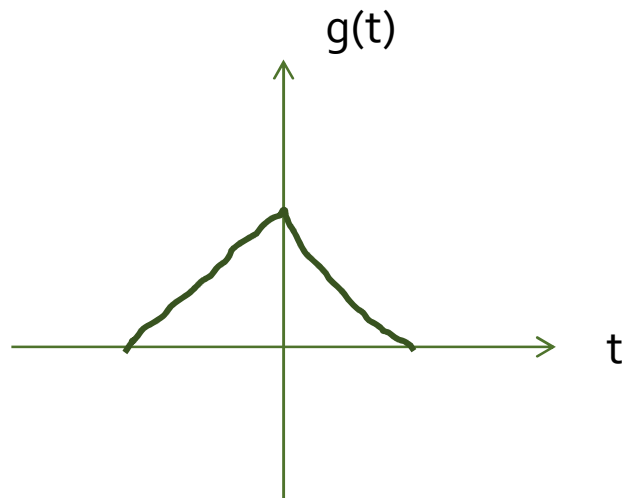


Properties: Linearity

- If $g_1(t) \Leftrightarrow G_1(f)$ and $g_2(t) \Leftrightarrow G_2(f)$
- Then $c_1g_1(t) + c_2g_2(t) \Leftrightarrow c_1G_1(f) + c_2G_2(f)$
- The spectrum of a weighted-sum signal is the weighted-sum of individual signal's spectrum.
- Nice!

Properties: Time scaling

- $g(t) \rightarrow g(at)$



Properties: Time scaling

- if $F[g(t)] = G(f)$,
- Then $F[g(at)] = \int_{-\infty}^{\infty} g(at) \exp(-j2\pi ft) dt$

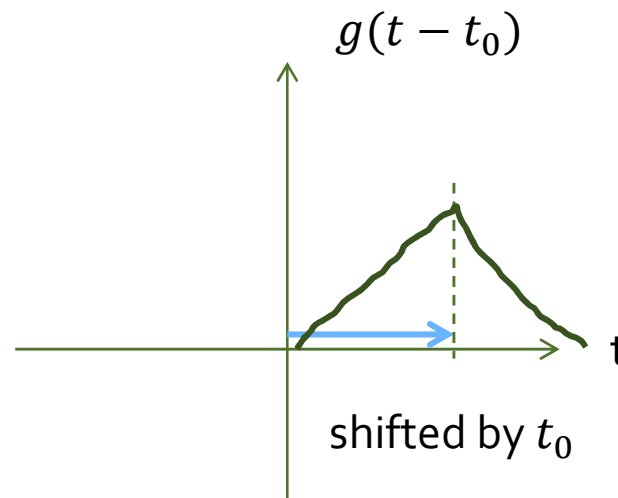
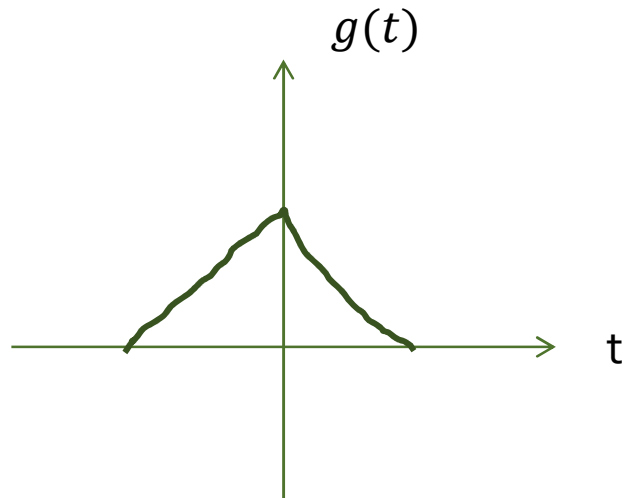
- set $\tau = at \Rightarrow d\tau = a dt$

- $$F[g(at)] = \frac{1}{|a|} \int_{-\infty}^{\infty} g(\tau) \exp\left(-j2\pi \frac{f}{a} \tau\right) d\tau$$
$$= \frac{1}{|a|} G\left(\frac{f}{a}\right)$$

Expanded in frequency by a factor of a !
("Faster signal" \rightarrow occupies larger bandwidth)

Properties: Time shifting

- $g(t) \rightarrow g(t - t_0)$



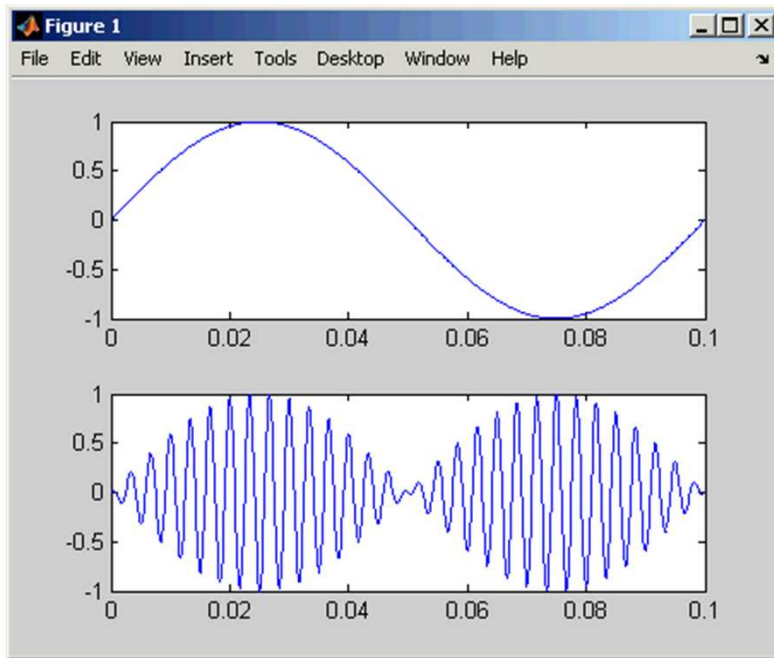
Properties: Time shifting

- if $F[g(t)] = G(f)$,
- then $F[g(t - t_0)] = \int_{-\infty}^{\infty} g(t - t_0) \exp(-j2\pi ft) dt$
- Set $\tau = t - t_0$, then $d\tau = dt$
- $F[g(t - t_0)] = \int_{-\infty}^{\infty} g(\tau) \exp(-j2\pi f\tau) \exp(-j2\pi ft_0) d\tau = \exp(-j2\pi ft_0) G(f)$
- **When the original signal is shifted in time:**
 - The amplitude of the Fourier transform is unchanged.
 - The phase is changed by the linear factor of $-j2\pi ft_0$.

Properties: Frequency shifting

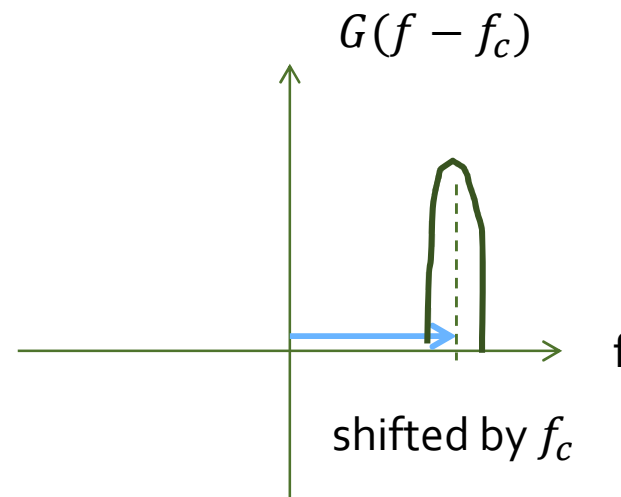
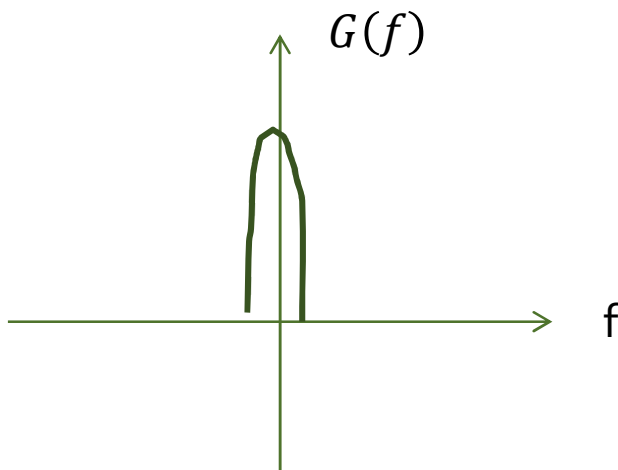
- if $F[g(t)] = G(f)$,
- $F[\exp(j2\pi f_c t)g(t)] =$
 $\int_{-\infty}^{\infty} g(t)\exp(j2\pi f_c t)\exp(-j2\pi ft)dt$
 $= \int_{-\infty}^{\infty} g(t)\exp(-j2\pi(f - f_c)t)dt$
 $= G(f - f_c)$
- This property is also called “modulation theorem”.
- Multiplication of a function by the factor $\exp(j2\pi f_c t)$ is equivalent to shifting its Fourier transform $G(f)$ in the positive direction by f_c .

Properties: Frequency shifting



Original Signal: $g(t)$

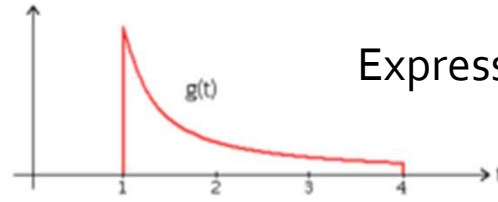
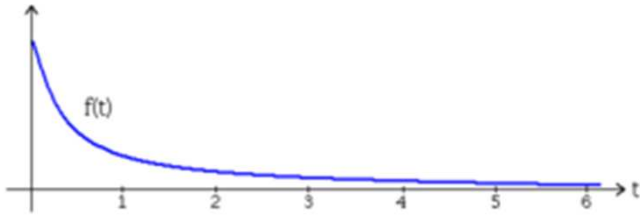
After modulated: $g(t)\exp(2\pi f_c t)$



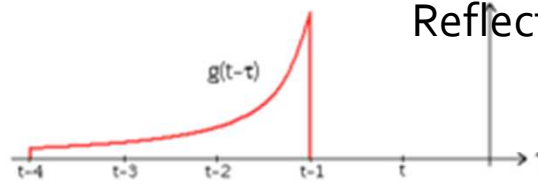
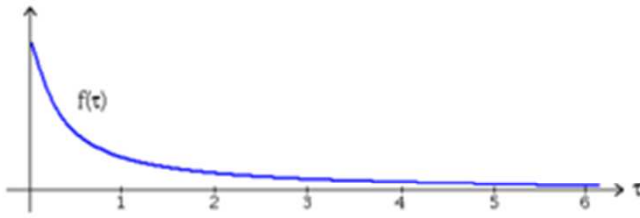
Convolution

- The convolution of f and g is written as $f * g$. It is a particular kind of integral transform:

$$\begin{aligned}(f * g)(t) &= \int_{-\infty}^{\infty} f(\tau)g(t - \tau)d\tau \\ &= \int_{-\infty}^{\infty} g(\tau)f(t - \tau)d\tau\end{aligned}$$

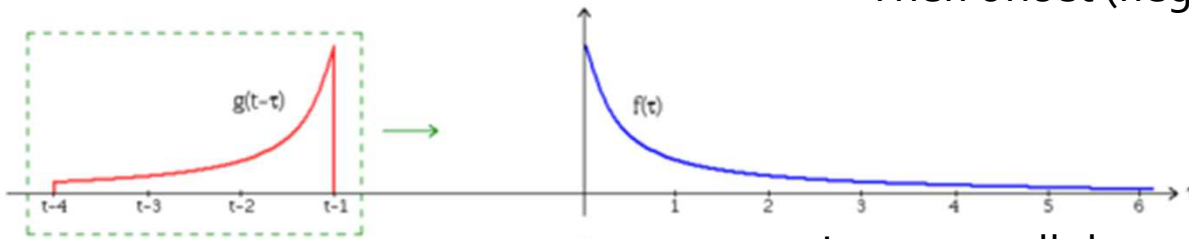


Express each function in terms of τ

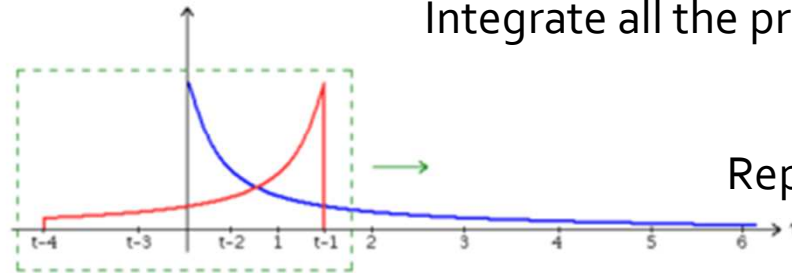


Reflect function g : $g(\tau) \rightarrow g(-\tau)$

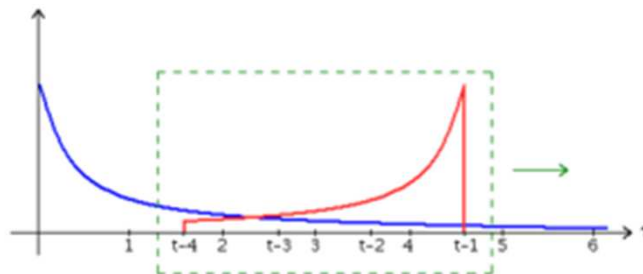
Then offset (negatively) by t : $g(-\tau) \rightarrow g(t - \tau)$



Integrate all the product of all intersect parts.



Repeat for a different t .



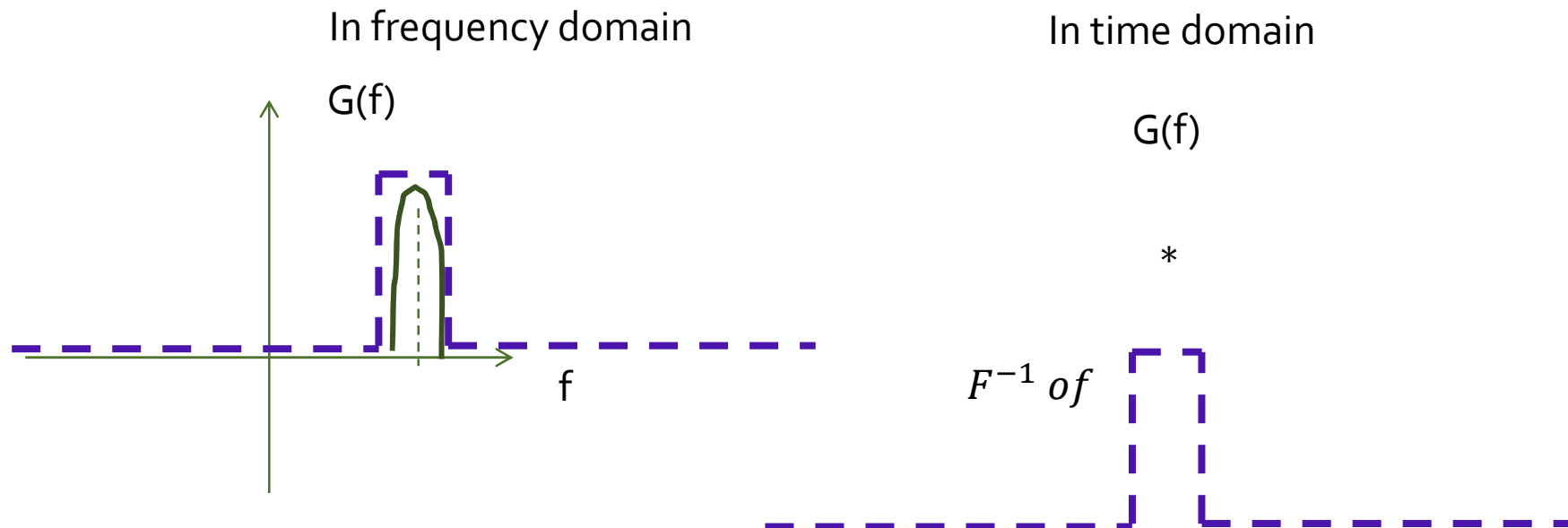
Properties: Multiplication in time Domain

- if $F[g_1(t)] = G_1(f)$ and $F[g_2(t)] = G_2(f)$,
- Then $F[g_1(t)g_2(t)] = G_{12}(f) = (G_1 * G_2)(f)$
- $G_{12}(f) = \int_{-\infty}^{\infty} g_1(t)g_2(t)\exp(-j2\pi ft)dt$
- $g_2(t) = \int_{-\infty}^{\infty} G_2(f')\exp(j2\pi f't)df'$
- $G_{12}(f) = \int_{-\infty}^{\infty} g_1(t) \int_{-\infty}^{\infty} G_2(f')\exp(j2\pi f't)df' \exp(-j2\pi ft)dt$
- Set $\lambda = f - f'$, $d\lambda = -df'$
- $G_{12}(f) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} g_1(t)G_2(f - \lambda)\exp(-j2\pi\lambda t)(-d\lambda)dt$
$$= \int_{-\infty}^{\infty} G_1(f)G_2(f - \lambda)d\lambda$$
$$= (G_1 * G_2)(f)$$

Properties: Convolution in time Domain

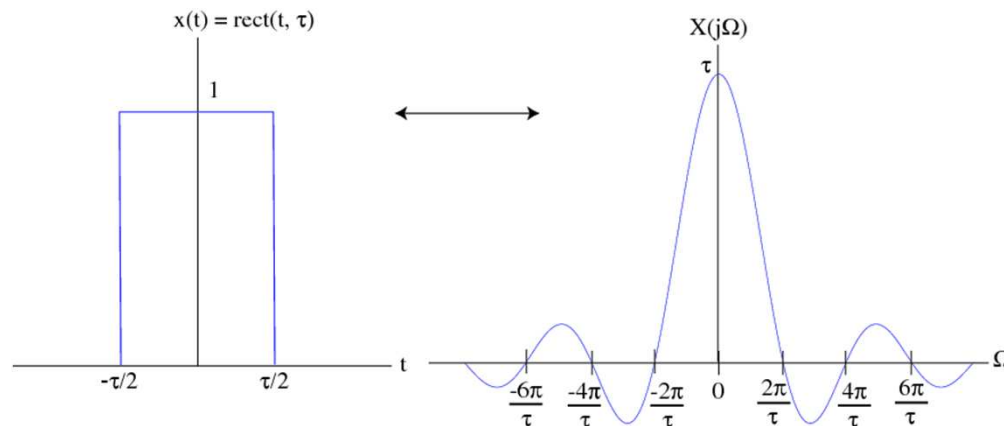
- if $F[g_1(t)] = G_1(f)$ and $F[g_2(t)] = G_2(f)$,
- Then $F[(g_1 * g_2)(t)] = G_1(f)G_2(f)$

- **Filter:**



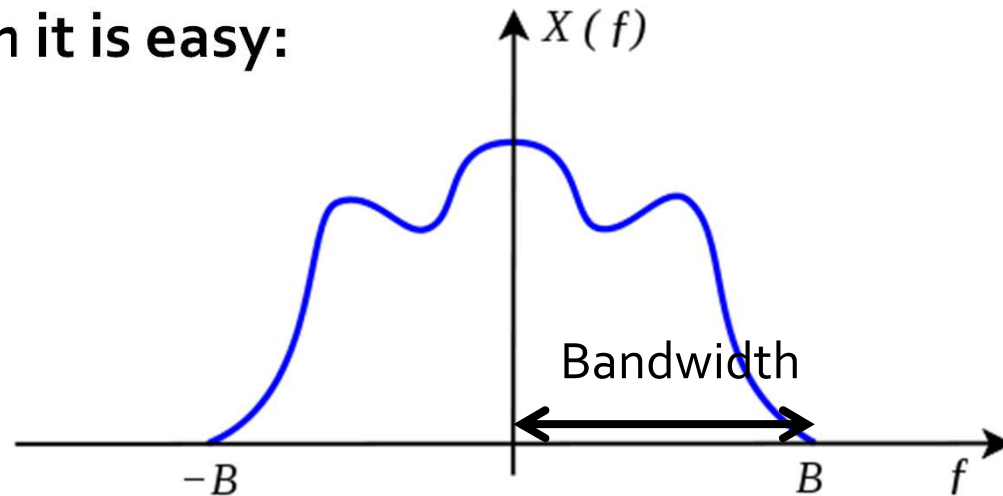
More properties

- **When the time domain signal is changed, the frequency domain signal is inversely changed.**
 - You can only specify one of the two, but not both at the same time.
- **X-limited: only a certain range is non-zero.**
 - Time-limited \rightarrow The signal trails on indefinitely in the frequency domain
 - Band-limited \rightarrow The signal trails on indefinitely in the time domain



Bandwidth

- The bandwidth of a signal provides a measure of the extent of significant spectral content of the signal for positive frequencies.
- But how “significant”?
- If band-limited, then it is easy:

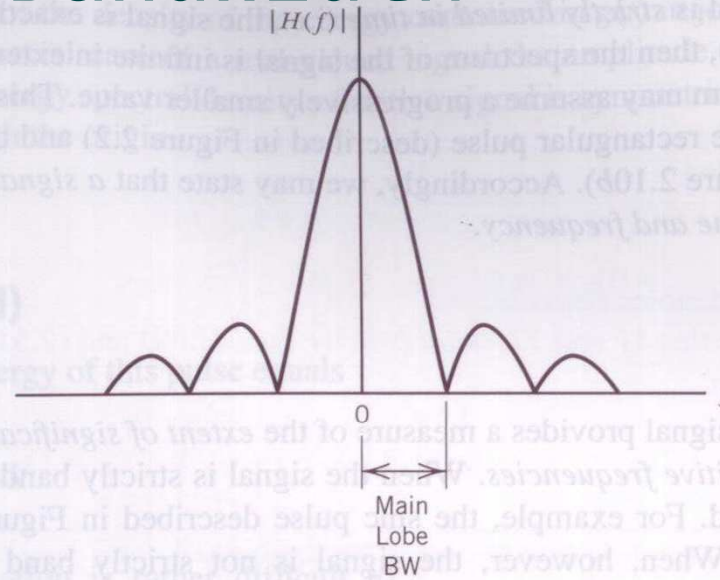


Bandwidth

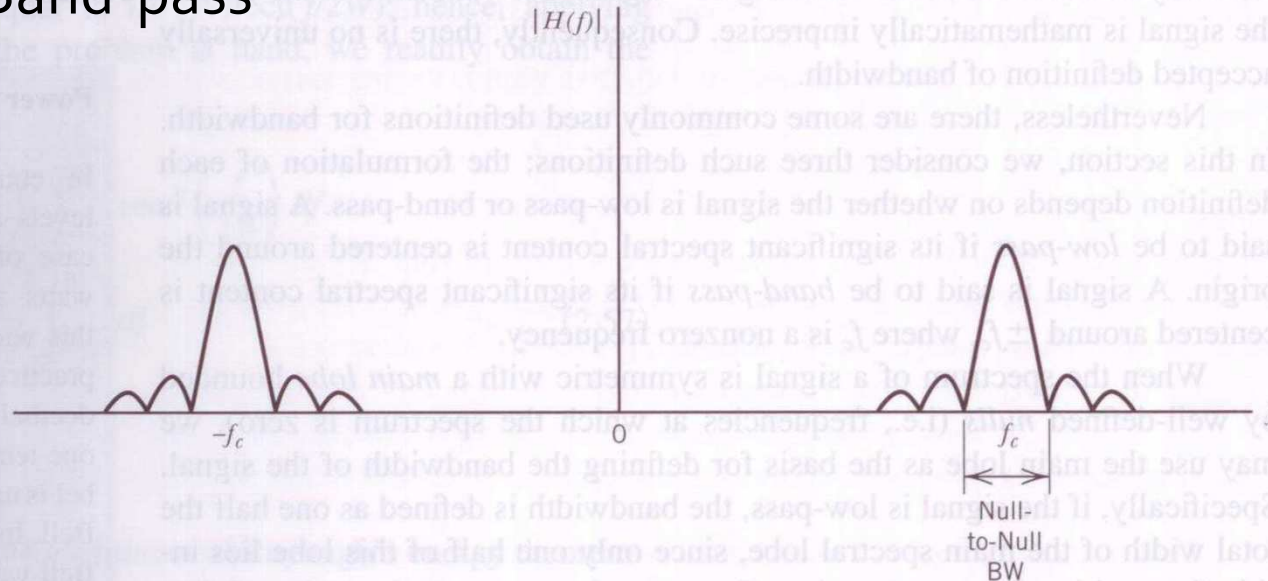
- **Low-pass signal:**
the signal's significant spectral content is centered around the origin.
- **Band-pass signal:**
the signal's significant spectral content is centered around $\pm f_c, f_c \neq 0$.
- **Main-lobe:**
Bounded by nulls (frequencies at which the spectrum = 0)

Main-lobe Bandwidth

Low-pass

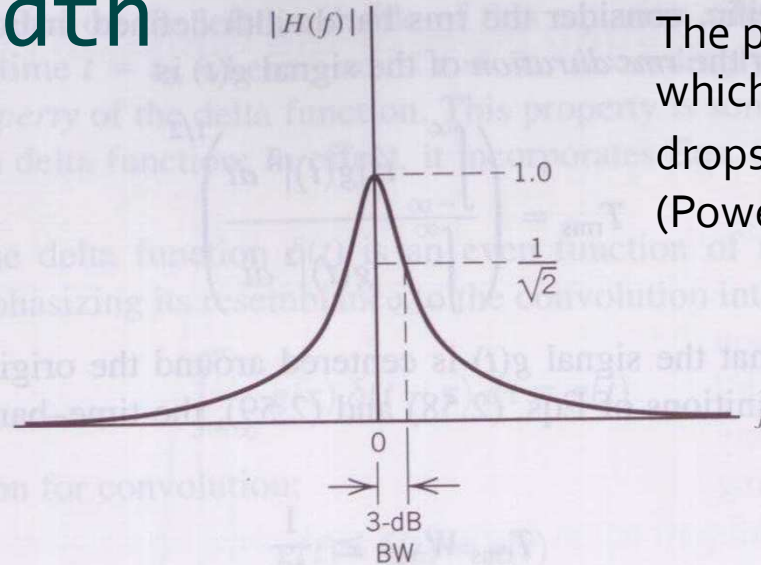


Band-pass



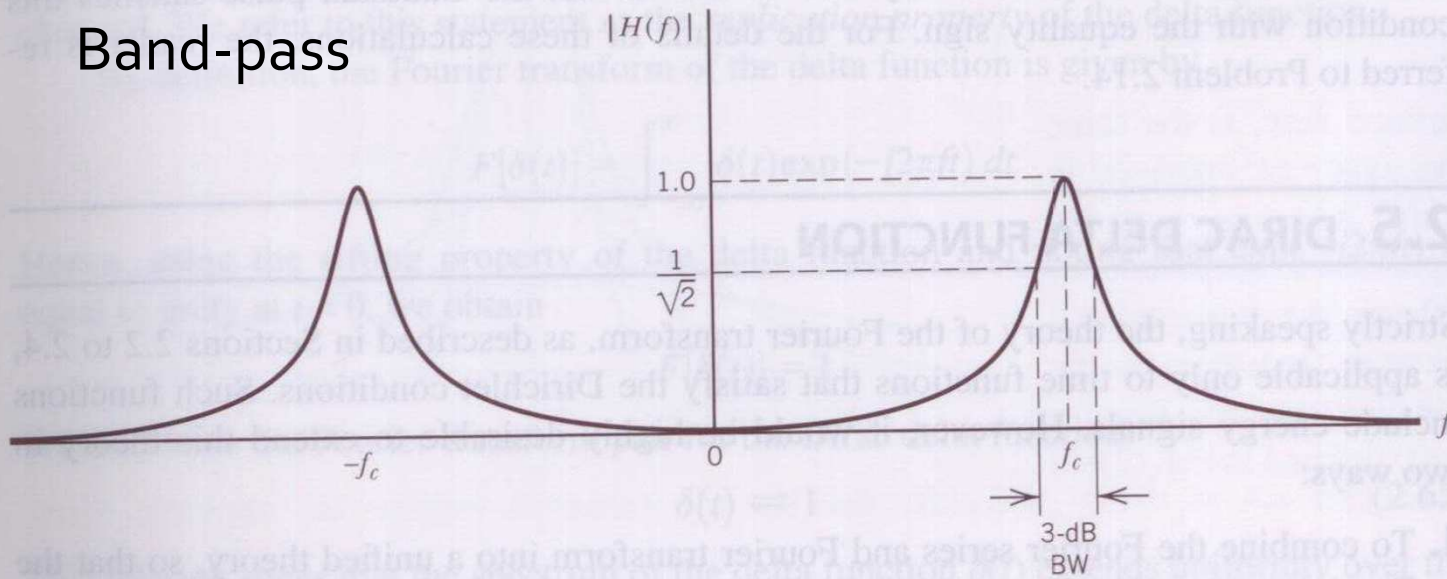
3-dB Bandwidth

Low-pass



The positive frequency at which the amplitude spectrum drops to $1/\sqrt{2}$ of the peak. (Power is $1/2$, or -3dB)

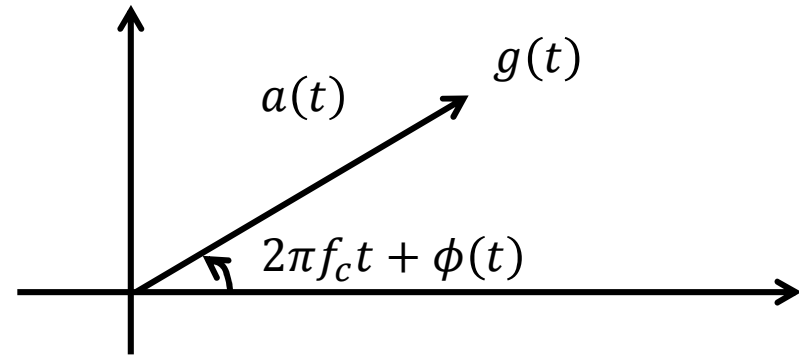
Band-pass



Not good for slow decaying spectrum.

Band-pass Signal Representation

- General form:



$$g(t) = a(t)\cos(2\pi f_c t + \phi(t))$$

Envelope

Phase

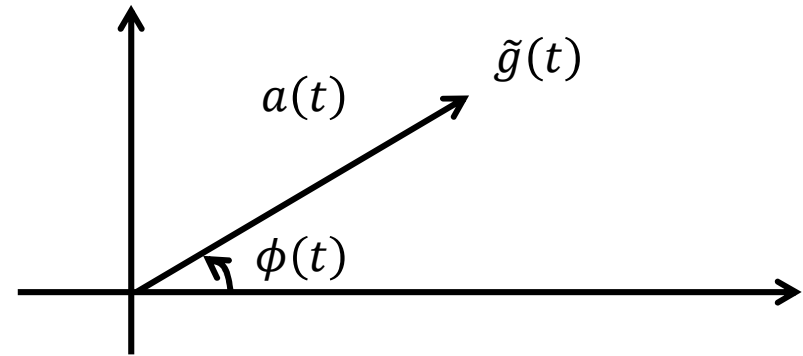
- Envelope is always non-negative, or we can switch the phase by 180 degree
- This is called the canonical representation of a band-pass signal

Band-pass Signal Representation

- $g(t) = a(t)\cos(2\pi f_c t + \phi(t))$ can be re-arranged into
- $g(t) = g_I(t)\cos(2\pi f_c t) - g_Q(t)\sin(2\pi f_c t)$
- $g_I(t) = a(t)\cos(\phi(t))$ and $g_Q(t) = a(t)\sin(\phi(t))$
- $g_I(t)$ and $g_Q(t)$ are called inphase and quadrature components of the signal $g(t)$, respectively
- Then $a(t) = \sqrt{g_I^2(t) + g_Q^2(t)}$ and $\phi(t) = \tan^{-1}\left(\frac{g_Q(t)}{g_I(t)}\right)$

Band-pass Signal Representation

- We can also represent $g(t)$ as



$$g(t) = \text{Re}[\tilde{g}(t)\exp(j2\pi f_c t)]$$

- $\tilde{g}(t) = g_I(t) + jg_Q(t)$
- $\tilde{g}(t)$ is called the complex envelope of the band-pass signal.
- This is to remove the annoying $\exp(j2\pi f_c t)$ in the analysis.